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THE DESIGN OF FUZZY LOGIC CONTROLLER VIA COPYING A LINEAR CONTROLLER

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Abstract

The design of the fuzzy logic controller via copying a Linear Quadratic Regulator (LQR) is presented. To synthesize a fuzzy controller, we pursued the idea of making it match the LQR for small inputs since the LQR was so successful^[9]. Then we still have the added tuning flexibility with the fuzzy controller to shape the control surface so that for larger inputs, it can perform differently from the LQR. The 25 "If-Then" rules determined heuristically based on the knowledge of the plant dynamics were stored in the MATLAB workspace from where they were transferred into the fuzzy controller model ready to be used for simulation in MATLAB/Simulink environment.

Keywords: Fuzzy controller, LQR, tuning flexibility, synthesized controller, If-Then rules.

1.0 INTRODUCTION

Fuzzy controller can be viewed as an artificial decision maker that operates in a closed loop system in real time (see fig. 1.1). It gathers the plant output data y(t), compares it with the reference input r(t) and decides on what the plant input u(t) should be to ensure the performance objectives are met. Note that every fuzzy control system is a nonlinear system and more so that even if the plant is linear, fuzzy and hence fuzzy control is nonlinear.

To design the fuzzy controller, the control engineer must gather enough information on how artificial decision maker should act in a closed-loop system. Sometimes, this information could come from a human decision maker (the machine operator), who performs the control task, while at other times, it may come from the control engineer who comes to understand the plant dynamics and write down a set of rules about how to control the system without outside help^[7].

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These "rules" basically say, "If the plant output and reference input are behaving in a certain manner then the plant input should be some value." A whole set of such "If-Then" rules are loaded into the rule base and inference strategy is chosen, then the system is ready to be tested to see if the closed-loop specifications are met.

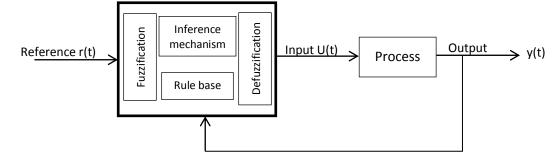


Fig. 1:1 Fuzzy Controller Architecture^{[9].}

1.1 Conventional/Fuzzy Control Approach to Balancing Control

There is no general systematic procedure for the design of fuzzy controllers that will definitely produce a high performance fuzzy control system for a wide variety of application. It is necessary to learn about fuzzy controller design via example^[9]. Although, numerous linear control design techniques have been applied to this particular system, here we consider the performance of only the LQR. The objective is of two folds:^[8]

- i. To form a base line for comparison of fuzzy control design to follow.
- ii. To provide a starting point for the synthesis of fuzzy controller.

A linear map such as LQR can be easily approximated by a fuzzy system (at least for small values of inputs to the fuzzy system). For the fact that a linearized system is completely controllable and observable, linear state-feedback strategies, such as LQR is applicable. The performance index for the LQR is given by^[5],

$$J = \int_{0}^{\infty} (X_{(t)}^{T} Q x_{(t)} + U_{(t)}^{T} R u_{(t)}) dt$$
(1.1)

Where Q and R are the weighting matrices of appropriate dimension corresponding to the state X and input U respectively. Given a fixed Q and R, the feedback gains that optimize the function J can be uniquely determined by solving an algebraic Riccati equation in MATLAB.

In effect, the idea of using the LQR controller is analogous to the actual implementation of results for the case where there is no additional load or mass to the end point (i.e. nominal case) in a pendulum. Now, when an extra weight or sloshing liquid (using a watertight bottle) is attached to the end point of the pendulum, for instance, the performance of all the linear

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controller degrades considerably, often resulting in unstable behaviour, and here a fuzzy control is needed^[7].

2.0 DESIGN

Assumptions:

The following assumptions were made during the design^[7]:

- i. Assume that all the input universes of discourse have uniformly distributed triangular membership functions (as shown in fig. 2.1) with the effective universes of discourse, all given by (-1, +1) i.e. so that the left-most membership function and the right-most membership function saturate at -1 and +1 respectively.
- ii. Assume we label the membership functions with linguistic numeric indices that are integers with zero at the middle.
- iii. Assume an isolated power station with the following parameters (see table 2.1).

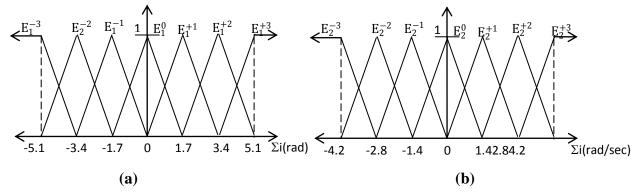


Fig. 2.1: Set of input membership functions^[9].

Gain		Time	e Const	ant
Turbine	$K_T = 1$	γ_r	=	0.5
Governor	$K_g = 1$	$\gamma_{\rm g}$	=	0.2
Amplifier	$K_A = 10$	γ _A	=	0.1
Exciter	$K_E = 1$	$\gamma_{\rm E}$	=	0.4
Generator	$K_{G} = 0.8$	γ _G	=	1.4
Sensor	$K_R = 1$	γ_R	=	0.05
Inertia	H = 5			
Regulation	R = 0.05			

 Table 2.1: Parameters for an Isolated Power Station^[5]

Assume a sudden load change of $\Delta P_L = 0.2$ (p.u).

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2.1 METHODOLOGY

2.1.1 Summation Operation

First, the IF – THEN rules were arranged so that the output membership function centres are equal to a scaled sum of the premise linguistic numeric indices. For a fuzzy controller in general, the centre of the controller output is given by^[9]:-

$$Y = (j + k ... + l)x \frac{2}{(N-1)n}$$
(2.1)

Where Y is the value of the output membership function centres (j + k ... + l) are sum of the premise linguistic numeric indices

N is the number of the membership functions in each input universe of discourse.

n is the number of inputs as assumed by the designer.

If for instance,

j + k = sum of the premise linguistic numeric indices N = 5 (as a standard) n = 2

Substituting the above values, equation 2.1 reduces to:

$$Y = (j + k) x \frac{2}{(5-1)^2}$$

$$= (j + k) \frac{1}{4}$$
(2.2)

Equation 2.2 was then used to generate the rule base shown in table 2.2

Output Centre		"Input 2" j index				
		-2	-1	0	1	2
"Input 1"	-2	-1	-0.75	-0.5	-0.25	0
K index	-1	-0.75	-0.5	-0.25	0	0.25
	0	-0.5	-0.25	0	0.25	0.5
	1	-0.25	0	0.25	0.5	0.75
	2	0	0.25	0.5	0.75	+1

 Table 2.2: Rule table created for copying a linear controller

Observations:

• From table 2.2, it can be seen that the fuzzy system is normalized i.e. the effective universes of discourse for the inputs and outputs are (-1, +1).

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- Again, the body of the table represents the centres of nine distinct output membership function centres.
- Note the diagonal of zeros; and by viewing the body of the table as a matrix, we see that it has a certain symmetry to it. This symmetry is not by accident but a representation of abstract knowledge about how to control the system^[7].

Table of Linguistic Variable with Rules:

The linguistic variables were written from the output membership function centres of table 2.2. Hence, we have table 2.3 as shown.

Table 2.3: Showing the Linguistic Variables (U_s = output membership function centres;

U _s		c				
		NB	NS	Ζ	PS	PB
	NB	NB	NB	NB	NS	Z
	NS	NB	NB	NS	Z	PS
e	Z	NB	NS	Ζ	PS	PB
	PS	NS	Z	PS	PB	PB
	PB	Z	PS	PB	PB	PB

e = error and c = change in error)

The output membership function centres are 25 in number, and these represent 25 IF-THEN rules from this table. These 25 rules were clearly spelt and stated, and then stored in the MATLAB workspace from where they were transferred into the fuzzy controller by opening its dialog box. Here are some of the IF-THEN rules organized from the rule table:-

IF e is negative big NB and c is negative big NB, THEN U_s is negative big NB. IF e is negative small NS and c is negative big NB THEN U_s is negative big NB. IF e is zero Z and c is negative big NB THEN U_s is negative big NB. IF e is positive small PS and c is negative big NB THEN U_s is negative small NS. IF e is positive big PB and c is negative big NB THEN U_s is zero Z, and so on.

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2.1.2 The Scaling Gain

Having achieved so far in the design, next is to pick the scaling gains so that the fuzzy system implements a weighted sum. Prior to this, the LQR was first designed by solving its algebraic Riccati equation in MATLAB^[5], thus:

Given performance index J = $\int_0^\infty (X_1^2 + 5x_3^2 + U^2) dt$ (2.3)

The state equation in matrix form^[5] is given by

$$\begin{bmatrix} \dot{\Delta P} \mathbf{v} \\ \Delta \dot{P} \mathbf{m} \\ \Delta \dot{\omega} \end{bmatrix} = \begin{bmatrix} \frac{-1}{\gamma_g} & 0 & \frac{-1}{R\gamma_g} \\ \frac{1}{\gamma_T} \frac{1}{\gamma_T} & 0 \\ 0 & \frac{1}{2H} \frac{-D}{2H} \end{bmatrix} \begin{bmatrix} \Delta P \mathbf{v} \\ \Delta P \mathbf{m} \\ \Delta \omega \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ \frac{-1}{2H} \end{bmatrix} \Delta P_L + \begin{bmatrix} \frac{1}{\gamma_g} \\ 0 \\ 0 \end{bmatrix} \Delta P_{ref}$$
(2.4)

Substituting the parameter values of the system (see the third assumption) into the state equation (2.4), with $\Delta P_{ref} = 0$ we have:

$$\mathbf{x} = \begin{bmatrix} -5 & 0 & -100 \\ 2 & -2 & 0 \\ 0 & 0.1 & -0.08 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 0 \\ 0 \\ -0.1 \end{bmatrix} \mathbf{U}$$
(2.5)

and the output equation $y = \begin{bmatrix} 0 & 0 & 1 \end{bmatrix} x$ where $y = \Delta \omega$ and $x = \begin{bmatrix} \Delta Pv \\ \Delta Pm \\ \Delta \omega \end{bmatrix}$

 \therefore For this system we have by inspection:

$$A = \begin{bmatrix} -5 & 0 & -100 \\ 2 & -2 & 0 \\ 0 & 0.1 & -0.08 \end{bmatrix}, \quad B = \begin{bmatrix} 0 \\ 0 \\ -0.1 \end{bmatrix}, \quad C = \begin{bmatrix} 0 & 0 & 1 \end{bmatrix}, \quad Q = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 5 \end{bmatrix} \text{ and } R = 1$$

To obtain the optimal feedback gain vector to minimize the given performance index (eqn. 2.3), we use the following, MATLAB commands^[8]:-

$$\begin{array}{rcl} P_L & = & 0.2; \\ A & = & [-5 \ 0 \ -100; \ 2 \ -2 \ 0; \ 0 \ 0.1 \ -0.08]; \\ B & = & [0; 0; -0.1]; \ BP_L = \ P_L * B; \\ C & = & [0 \ 0 \ 1]; \ D = 0; \end{array}$$

(2.6)

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The result is

K₁, K₂, K₃ respectively

= 0.0834, -0.2875, -12.1381

Р

=	0.1100	0.0267	-0.8338
	0.0267	0.1231	2.8751
	-0.8338	2.8751	121.3810

 $A_{\rm f} \\$

=

-5.0000	0	-100.0000
2.0000	-2.0000	0
0.0083	0.0712	-1.2938

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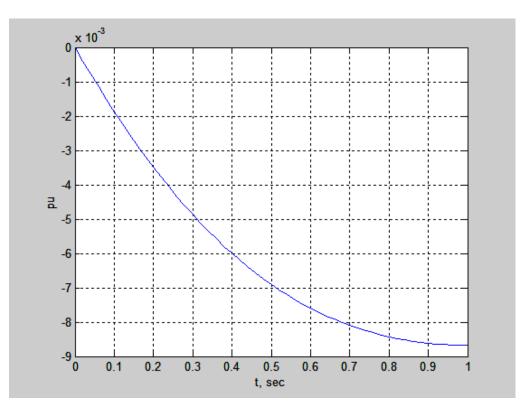


Fig. 2.2: Plot of frequency deviation responses

The simulation block diagram of the system condition is constructed as shown in fig. 2.3

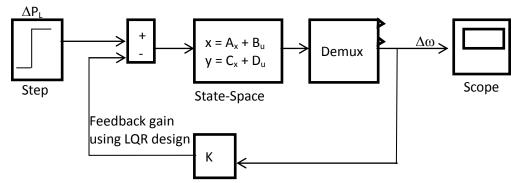


Fig. 2.3: Simulation Block Diagram of the System^[5].

The state-space description dialog box was opened and the values of A, B, C and D constants were entered in the appropriate box in Matlab matrix notation.

2.1.3 Choice of Scaling Gains

To choose the scaling gains so that the fuzzy system implements a weighted sum, the appropriate gain on the ith input-output pair is g_ig_o ; so to copy the ki of the state feedback controller, we choose^[9]:-

 $\begin{array}{rcl} g_i g_o &=& ki \end{array} (2.7) \\ \label{eq:gi} Where ki &=& optimal gain of LQR \\ g_i &=& i^{th} input gain \\ g_o &=& i^{th} output gain \end{array}$

Choosing the controller input that most greatly influences the plant behaviour to be $g_3 = 2^{[9]}(2.8)$

From the results of the optimal gains calculated, $K_3 = -12.14$.

 $g_{3}g_{0} = K_{3}$ from equation (2.7) and hence, $\frac{K_{3}}{g_{3}} = g_{0}$ $\therefore g_{0} = \frac{-12.1381}{2} = -6.069$

Now, for $g_0 = -6.07$, the input gains are:

$$g_1 = \frac{0.0834}{-6.07} = -0.01374$$

$$g_2 = \frac{-0.2875}{-6.07} = 0.04736$$

$$g_3 = \frac{-12.138}{-6.07} = 1.9996 \cong 2$$

Try the scaling gains of 0.04736 or 2.

3.0 ANALYSIS RESULT

The gains of 2 were chosen so that the fuzzy system implements a weighted sum and is inserted in any model bearing the fuzzy controller and any chosen compensator. A simulation is done in Matlab/Simulink environment to get a result.

4.0 CONCLUSION

Fuzzy logic is dynamic because human expert has the responsibility to control the plant; and fuzzy system is very important because it not only helps to reduce complex mathematical

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model in problem solving, it is a natural language and easy to understand. Above all, it acts as a supplementary controller to major controllers in a given system.

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